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# Secrecy Spectrum and Secrecy Energy Efficiency in Massive MIMO Enabled HetNets

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#### Abstract

Security and resource-saving are both demands of the fifth generation (5G) wireless networks. In this paper, we study the secrecy spectrum efficiency (SSE) and secrecy energy efficiency (SEE) of a *K*-tier massive multiple-input multiple-output (MIMO) enabled heterogeneous cellular network (HetNet), in which artificial noise (AN) are employed for secrecy enhancement. Assuming (i) independent Poisson point process model for the locations of base stations (BSs) of each tier as well as that of eavesdroppers, (ii) zero-forcing precoding at the macrocell BSs (MBSs), and (iii) maximum average received power-based cell selection, the tractable lower bound expressions for SSE and SEE of massive MIMO enabled HetNets are derived. Then, the influences on secrecy oriented spectrum and energy efficiency performance caused by the power allocation for AN, transmit antenna number, number of users served by each MBS, and eavesdropper density are analyzed respectively. Moreover, the analysis accuracy is verified by Monte Carlo simulations.

**Keywords:** Heterogeneous cellular networks, physical layer security, massive MIMO, energy efficiency, artificial noise

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## **1. Introduction**

Owing to the ever-growing demand for high-quality high-rate wireless communications, the fifth generation (5G) cellular networks in the future are likely to be massive multiple-input multiple-output (MIMO) enabled heterogeneous cellular networks (HetNets) [1]. With a mixture of macrocells and larger number of smaller cells such as microcells, picocells, and femtocells, HetNets bring the network closer to end users, thereby providing higher throughput and coverage [2]. Massive MIMO, also named large-scale antenna arrays, is an emerging physical layer technology which can produce high-resolution spatial beam and significantly improve radio spectral efficiency in HetNets [3]. These benefits have put massive MIMO enabled HetNets in the spotlight of preliminary 5G discussions [4].

In the past few years, the theoretical understanding of massive MIMO enabled HetNets has progressed significantly [5]. Recently, stochastic geometry is emerging as a popular technique that supports comprehensive yet tractable analysis of massive MIMO enabled HetNets [6-7]. To improve throughput in massive MIMO enabled HetNets, inter-tier interference coordination strategies were proposed in [8]. User association and interference management in massive MIMO aided HetNets were further investigated in [9] and [10]. Spectrum and energy efficiency in massive MIMO enabled HetNets were analyzed in [7] and [11].

Since the open nature of wireless channels, ensuring security is an essential task for wireless communication systems regardless of the type of network structure utilized. Physical layer security (PLS) has drawn ever-increasing attention as it can realize secure communication without using any ciphering keys [12]. Research efforts on the PLS have been made by considering different aspects, such as artificial noise (AN) [13], beamforming [14], and cooperative jamming [15], etc. Very recently, the investigation of PLS in HetNets was pioneered by the authors of [16]. Based on stochastic geometry, secrecy outage probability and network-wide secrecy throughput for HetNets were examined in [17] and [18]. PLS for massive MIMO aided HetNets was first investigated in [19]. A very recent contribution on coordinated multipoint secrecy transmission in HetNets was made by the authors of [20].

While there is a strong desire to improve the coverage and security on the one hand, the industry is on the other hand constantly facing the increasing challenge of meeting the demand of green communications [21]. Throughput and secrecy capacity maximization leads to large spectrum and energy consumptions, which are not desirable in forthcoming communication networks. In order to promote economic and ecological sustainability, nowadays researchers have made efforts to study the energy and spectral efficiency tradeoff instead of traditional pursuit of higher capacity and coverage [22-23]. Nevertheless, secrecy oriented spectrum and energy efficiency performance has been rarely discussed in the above researches. The researches on the energy and spectral efficiency problem of massive MIMO enabled HetNets, such as [7] and [11], mainly focused on the scenario without the existence of eavesdroppers. While many security issues in HetNets have been studied in the literature [12-20], controlling the spectrum and energy consumption on securing mobile communications in massive MIMO enabled HetNets has not been actively addressed. To the best of our knowledge, no prior work has accounted for secrecy spectrum efficiency (SSE) and secrecy energy efficiency (SEE) when designing massive MIMO enabled HetNets, and a fundamental analysis framework to evaluate the secrecy oriented spectrum and energy efficiency performance in massive MIMO enabled HetNets is lacking. Motivated by the above reasons as well as the demands for security and resource-saving in the future wireless networks, we aim to propose a tractable

approach for secrecy spectrum efficiency (SSE) and secrecy energy efficiency (SEE) analysis in the downlink of a *K*-tier massive MIMO enabled HetNets.

For clarity, our main contributions are summarized below:

- The achievable ergodic rate for macrocell base station (MBS) users, small-cell base station (SBS) users, and eavesdroppers of *K*-tier massive MIMO enabled HetNets are derived respectively on the condition that each MBS transmits artificial noise (AN) in the null space of the main channel to guarantee the secrecy at the intended receiver.
- Under the same conditions, a fundamental analysis framework to evaluate the secrecy oriented spectrum and energy efficiency performance in massive MIMO enabled HetNets is proposed.
- The impacts of AN power allocation ratio, MBS's transmit antenna number, number of users served by each MBS, and eavesdropper density on SSE as well as SEE are examined.

The remainder of this paper is organized as follows: Section 2 outlines the system model and the fundamental expressions. Section 3 introduces the derivations and analyses of SSE and SEE. Section 4 presents the numerical results. Finally, Section 5 concludes this study.

Notations:  $|\cdot|$ ,  $P\{\cdot\}$ , and  $E_A\{\cdot\}$  denote absolute value, probability, and expectation with respect to A, respectively.  $\Gamma\{N,\lambda\}$  stands for gamma distribution with parameters N and  $\lambda$ .  $\exp(\lambda)$  represents exponential distribution with parameter  $\lambda \cdot f_A(\cdot)$  and  $F_A(\cdot)$  denote probability density function and cumulative distribution function (CDF) of a random variable A, respectively.  $\Gamma(z)$  represents the gamma function with parameters  $z \cdot B_{(x)}[p,q]$  represents the incomplete beta function [24, (8.391)] with parameters x, p, and  $q \cdot {}_2F_1[\alpha,\beta;\gamma;z]$  is the Gauss hypergeometric function [24, (9.142)] with parameters  $\alpha$ ,  $\beta$ ,  $\gamma$ , and  $z \cdot L_A$  stands for Laplace transform with respect to A. Additionally,  $\{x\}^+ = \max\{x, 0\}$ .

# 2. System Model

In this study, we consider a *K*-tier HetNets consisting of macrocells and small-cells such as picocells and femtocells. As in other studies, we assume that the locations of the base stations (BSs) in the *i*<sup>th</sup> tier and eavesdroppers are characterized by two independent homogeneous Poisson point process (PPP)  $\Phi_i$  and  $\Phi_E$  with densities  $\lambda_i$  and  $\lambda_E$ , respectively. The first tier represents the class of MBSs, where each *N*-antenna MBS adopting massive MIMO simultaneously communicates with *S* users ( $N >> S \ge 1$ ). SBSs, legitimate users, and eavesdroppers are each equipped with a single antenna. In the *i*<sup>th</sup> tier, total transmit power of each BS is  $P_i$ , the path loss exponent is  $\alpha_i$ , and the biasing factor is  $B_i$ ,  $i \in [2, K]$ . Under the assumption of co-channel deployment and perfect channel state information, each MBS uses zero-forcing beamforming to transmit *S* data streams with equal power assignment. Meanwhile, each MBS transmits artificial noise (AN) in the null space of the main channel to guarantee the secrecy at the intended receiver. All channels in the system are assumed to undergo independent and identically distributed quasi-static Rayleigh fading.

#### 2.1 Legitimate Users

Without loss of generality, we analyze a typical users located at the origin o and consider a mobile association policy that associates each legitimate user with the BS providing the

maximum average received signal power. Hence, the receive signal to interference plus noise ratio (SINR) of a typical user at a random distance  $|X_{M,o}|$  from its associated MBS  $o_M$  can be expressed as

$$\operatorname{SINR}_{M} = \frac{(1-\eta)P_{1}h_{M,o}\beta \left|X_{M,o}\right|^{-\alpha_{1}}}{S(I_{M}+\delta^{2})},$$
(1)

where  $\eta \in [0,1]$  denotes the power allocation ratio of the AN power to the total transmit power,  $h_{M,o} \sim \Gamma(N-S+1,1)$  is the small-scale fading channel power gain between  $o_M$  and the typical user,  $\beta$  is the frequency dependent constant value,  $\delta^2$  is the additive noise power, and  $I_M = I_{M,M} + I_{M,S} + I_{M,AN}$ . The intra-tier interference for a MBS user is given by

$$I_{M,M} = \sum_{l \in \Phi_1 \setminus o_M} \frac{(1-\eta) P_1}{S} h_{M,l} \beta |X_{M,l}|^{-\alpha_1},$$
(2)

where  $h_{M,l} \sim \Gamma(S,1)$  and  $|X_{M,l}|$  are the equivalent small-scale fading interfering channel power gain and distance between the typical user and MBS  $l \in \Phi_1 \setminus o_M$  (except the serving MBS  $o_M$ ), respectively. The cross-tier interference for a MBS user can be written as

$$I_{M,S} = \sum_{i=2}^{K} \sum_{j \in \Phi_i} P_i h_{i,j} \beta \left| X_{i,j} \right|^{-\alpha_i},$$
(3)

where  $h_{i,j} \sim \exp(1)$  and  $|X_{i,j}|$  are the small-scale fading interfering channel power gain and distance between the typical user and SBS *j* in the *i*<sup>th</sup> tier, respectively. The artificial noise interference received by a MBS user is given by

$$I_{M,AN} = \sum_{l \in \Phi_1 \setminus o_M} \frac{\eta P_1}{S} h_{AN,l} \beta \left| X_{M,l} \right|^{-\alpha_1} , \qquad (4)$$

where  $h_{AN,l} \sim \Gamma(S,1)$  is the equivalent AN interfering channel power gain between the typical user and MBS *l*.

The SINR of a typical user at a random distance  $|X_{k,o}|$  from its associated SBS  $o_k$  in the  $k^{\text{th}}$  tier ( $k \in [2, K]$ ) can be formulated as

$$\operatorname{SINR}_{k} = \frac{P_{k} g_{k,o} \beta \left| X_{k,o} \right|^{-\alpha_{k}}}{I_{k} + \delta^{2}}, \qquad (5)$$

where  $g_{k,o} \sim \exp(1)$  is the small-scale fading channel power gain between the typical user and its serving SBS  $o_k$ , and  $I_k = I_{k,M} + I_{k,S} + I_{k,AN}$ . The macrocells-tier interference for a SBS user is given by

$$I_{k,M} = \sum_{l \in \Phi_1} \frac{(1-\eta)P_1}{S} g_{M,l} \beta |X_{M,l}|^{-\alpha_1} , \qquad (6)$$

where  $g_{M,l} \sim \Gamma(S,1)$  is the equivalent small-scale fading interfering channel power gain between the typical user and MBS *l*. The small-cells-tier interference for a SBS user can be written as

$$I_{k,S} = \sum_{i=2}^{K} \sum_{j \in \Phi_i \setminus o_k} P_i g_{i,j} \beta \left| X_{i,j} \right|^{-\alpha_i},$$
(7)

where  $g_{i,j} \sim \exp(1)$  is the small-scale fading interfering channel power gain between the typical user and SBS *j* in the *i*<sup>th</sup> tier. The artificial noise interference received by a SBS user is given by

$$I_{k,AN} = \sum_{l \in \Phi_1 \setminus o_M} \frac{\eta P_1}{S} g_{AN,l} \beta \left| X_{M,l} \right|^{-\alpha_1}, \qquad (8)$$

where  $g_{AN,l} \sim \Gamma(S,1)$  is the equivalent AN interfering channel power gain between the typical user and MBS *l*.

# 2.2 Eavesdroppers

In order to analytically characterize the secrecy performance in massive MIMO enabled HetNets, the capability of not only legitimate users but also eavesdroppers needs to be considered. In this work, we consider a worst-case wiretap scenario in which eavesdroppers are assumed to have the capability of multiuser decoding and jointly wiretap the typical user [18]. Thus, most of the interference created by concurrent transmission of information signals can be completely resolved, and the rest of it can be incorporated into the constant noise. In other words, to an arbitrary eavesdropper, the non-ignorable interference is generated by AN.

When the typical user is connected with a MBS, the receive SINR at the most detrimental eavesdropper is given by

$$\operatorname{SINR}_{M}^{e^{*}} = \max_{e \in \Phi_{E}} \left\{ \frac{(1-\eta) P_{1} h_{M,e} \beta \left| X_{M,e} \right|^{-\alpha_{1}}}{S(I_{M,e} + \delta^{2})} \right\},$$
(9)

where  $h_{M,e} \sim \exp(1)$  and  $|X_{M,e}|$  are the equivalent small-scale fading channel power gain and distance between the eavesdropper and its targeted MBS, respectively. The aggregate interference for a MBS eavesdropper can be written as

$$I_{M,e} = \sum_{l \in \Phi_1 \setminus o_M} \frac{\eta P_1}{S} h_{e,l} \beta |X_{e,l}|^{-\alpha_1} , \qquad (10)$$

where  $h_{e,l} \sim \Gamma(S,1)$  and  $|X_{e,l}|$  are the equivalent AN interfering channel power gain and distance between the eavesdropper and MBS *l*, respectively.

When the typical user is connected with a SBS in the  $k^{th}$  tier, the receive SINR at the most detrimental eavesdropper can be determined with

$$\operatorname{SINR}_{k}^{e^{*}} = \max_{e \in \Phi_{E}} \left\{ \frac{P_{k} g_{k,e} \beta \left| X_{k,e} \right|^{-\alpha_{k}}}{I_{k,e} + \delta^{2}} \right\},$$
(11)

where  $g_{k,e} \sim \exp(1)$  and  $|X_{k,e}|$  are the equivalent small-scale fading channel power gain and distance between the eavesdropper and its targeted SBS in the  $k^{\text{th}}$  tier, respectively. The aggregate interference for a SBS eavesdropper is given by

$$I_{k,e} = \sum_{l \in \Phi_1 \setminus o_M} \frac{\eta P_1}{S} g_{e,l} \beta \left| X_{e,l} \right|^{-\alpha_1}, \qquad (12)$$

where  $g_{e,l} \sim \Gamma(S,1)$  and  $|X_{e,l}|$  are the equivalent AN interfering channel power gain and distance between the eavesdropper and MBS *l*, respectively.

## 3. Secrecy Spectrum Efficiency and Secrecy Energy Efficiency

In this section, we investigate the network-wide spectrum efficiency and energy efficiency of the massive MIMO enabled HetNets based on the secrecy transmission capacity constraint, which is defined as the achievable rate of successful transmission of secret messages, i.e., achievable secrecy rate [25-26]. In an effort to assess the achievable secrecy rate, we first derive the achievable ergodic rate of the typical user and the most detrimental eavesdropper.

#### 3.1 Achievable ergodic rate

Given that an accurate achievable ergodic rate of the typical macrocell user is difficult to obtain even for traditional HetNets, we first derive the lower bounds of achievable ergodic rate for the typical macrocell user in massive MIMO enabled HetNets.

Lemma 1. The lower bound on the achievable ergodic rate for a randomly selected macrocell user can be expressed as

$$R_{M}^{L} = \log_{2} \left( 1 + \frac{(1-\eta)P_{1}(N-S+1)\beta}{S\mu} \right),$$
(13)

where

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$$\mu = \int_{0}^{\infty} x^{\alpha_{1}} \left( \mathbf{E} \{ I_{M} \} + \delta^{2} \right) f_{|X_{M,o}|}(x) \mathrm{d}x$$
(14)

with

$$E\{I_{M}\} = \frac{2\pi\lambda_{1}P_{1}\beta}{\alpha_{1}-2}x^{2-\alpha_{1}} + \sum_{i=2}^{K}\frac{2\pi\lambda_{i}P_{i}\beta}{\alpha_{i}-2}\left(\frac{P_{i}B_{i}Sx^{\alpha_{1}}}{(N-S+1)P_{1}}\right)^{\frac{2-\alpha_{i}}{\alpha_{i}}},$$
(15)

$$f_{|X_{M,o}|}(x) = \frac{2\pi\lambda_1}{A_M} x \exp\left\{-\pi\lambda_1 x^2 - \pi\sum_{i=2}^K \lambda_i \left(\frac{P_i B_i S x^{\alpha_1}}{(1-\eta) P_1(N-S+1)}\right)^{2/\alpha_i}\right\},$$
(16)

and

$$A_{M} = 2\pi\lambda_{1} \int_{0}^{\infty} r \exp\left\{-\pi\lambda_{1} r^{2} - \pi \sum_{i=2}^{K} \lambda_{i} \left(\frac{P_{i} B_{i} S r^{\alpha_{1}}}{(1-\eta) P_{1} (N-S+1)}\right)^{2/\alpha_{i}}\right\} \mathrm{d}r.$$
(17)

Here,  $A_M$  is the probability that a typical user is associated with a MBS.

**Proof.** See Appendix A.

**Lemma 2**. When the typical user is connected with a MBS, the achievable ergodic rate for the most detrimental eavesdropper can be given by

$$R_{M}^{e^{*}} = \frac{1}{\ln 2} \int_{0}^{\infty} \frac{1 - F_{SINR_{M}^{e^{*}}}(\gamma)}{1 + \gamma} d\gamma, \qquad (18)$$

where

$$\mathbf{F}_{\mathrm{SINR}_{M}^{*}}(\gamma) = \exp\left(-2\pi\lambda_{E}\int_{0}^{\infty}L_{I_{M,e}}\exp\left(-\frac{\gamma S\delta^{2}x^{\alpha_{1}}}{(1-\eta)P_{1}\beta}\right)xdx\right)$$
(19)

with

$$L_{I_{M,e}} = \exp\left(-2\pi\lambda_1 \sum_{n=1}^{S} {S \choose n} \frac{\left(\frac{\gamma\eta x^{\alpha_1}}{1-\eta}\right)^{\frac{2}{\alpha_1}} \Gamma\left(n-\frac{2}{\alpha_1}\right) \Gamma\left(-n+\frac{2}{\alpha_1}+S\right)}{\alpha_1 \Gamma\left(S\right)}\right).$$
(20)

**Proof.** See Appendix B.

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**Lemma 3**. For a randomly selected user associated with a SBS in the  $k^{\text{th}}$  tier, the achievable ergodic rate can be formulated as

$$R_{k} = \frac{1}{\ln 2} \int_{0}^{\infty} \frac{1 - F_{\text{SINR}_{k}}(\gamma)}{1 + \gamma} d\gamma, \qquad (21)$$

where

$$F_{SINR_{k}}(\gamma) = 1 - \int_{0}^{\infty} L_{I_{k}} \exp\left(-\frac{\gamma x^{\alpha_{k}} \delta^{2}}{P_{k} \beta}\right) f_{|X_{k,o}|}(x) dx$$
(22)

with

$$L_{I_{k}} = \exp\left\{-2\pi\lambda_{1}\sum_{\nu=1}^{S}\binom{S}{\nu}\left(\frac{\eta P_{1}\gamma x^{\alpha_{k}}}{SP_{k}}\right)^{\frac{2}{\alpha_{1}}}B_{\left(-\frac{\eta P_{1}\gamma x^{\alpha_{k}}}{SP_{k}}\left(D_{M}^{k}(x)\right)^{-\alpha_{1}}\right)}\left[\nu - \frac{2}{\alpha_{1}}, 1 - S\right]\right) + \left(\frac{(1 - \eta)P_{1}\gamma x^{\alpha_{k}}}{SP_{k}}\right)^{\frac{2}{\alpha_{1}}}B_{\left(\frac{(\eta - 1)P_{1}\gamma x^{\alpha_{k}}}{SP_{k}}\left(D_{M}^{k}(x)\right)^{-\alpha_{1}}\right)}\left[\nu - \frac{2}{\alpha_{1}}, 1 - S\right]\right)\left(-1\right)^{\nu}\frac{1}{\alpha_{1}}$$

$$-\sum_{i=2}^{K}\frac{2\pi\lambda_{i}P_{i}\gamma x^{\alpha_{k}}}{P_{k}(\alpha_{i} - 2)}\left(D_{i}^{k}(x)\right)^{2-\alpha_{i}} {}_{2}F_{1}\left[\frac{\alpha_{i} - 2}{\alpha_{i}}, 1; 2 - \frac{2}{\alpha_{i}}; -\frac{\gamma x^{\alpha_{k}}P_{i}}{P_{k}\left(D_{i}^{k}(x)\right)^{\alpha_{i}}}\right]\right\},$$
(23)

$$D_{M}^{k}(x) = \left(\frac{P_{1}(N-S+1)x^{\alpha_{k}}}{P_{k}B_{k}S}\right)^{\frac{1}{\alpha_{1}}},$$
(24)

$$D_i^k\left(x\right) = \left(\frac{P_i B_i x^{\alpha_k}}{P_k B_k}\right)^{\frac{1}{\alpha_1}},\tag{25}$$

$$f_{|X_{k,o}|}(x) = \frac{2\pi\lambda_k x}{A_k} \exp\left\{-\pi\lambda_1 \left(\frac{(1-\eta)P_1(N-S+1)x^{\alpha_k}}{P_i B_i S}\right)^{\frac{2}{\alpha_1}} - \pi\sum_{i=2}^K \lambda_i \left(\frac{P_i B_i x^{\alpha_k}}{P_k B_k}\right)^{\frac{2}{\alpha_i}}\right\}, \quad (26)$$

and

$$A_{k} = 2\pi\lambda_{k} \int_{0}^{\infty} r \exp\left\{-\pi\lambda_{1} \left(\frac{(1-\eta)P_{1}(N-S+1)r^{\alpha_{k}}}{P_{k}B_{k}S}\right)^{\frac{2}{\alpha_{1}}} - \pi\sum_{i=2}^{K}\lambda_{i} \left(\frac{P_{i}B_{i}r^{\alpha_{k}}}{P_{k}B_{k}}\right)^{\frac{2}{\alpha_{i}}}\right\} dr.$$
(27)

**Proof.** The proof of Lemma 3 follows similar lines of the proof in Lemma 1 and Lemma 2. Lemma 4. When the typical user is connected with a SBS in the  $k^{th}$  tier, the achievable ergodic rate for the most detrimental eavesdropper can be expressed as

$$R_{k}^{e^{*}} = \frac{1}{\ln 2} \int_{0}^{\infty} \frac{1 - F_{SINR_{k}^{e^{*}}}(\gamma)}{1 + \gamma} d\gamma , \qquad (28)$$

where

$$\mathbf{F}_{\mathrm{SINR}_{k}^{e^{*}}}(\gamma) = \exp\left(-2\pi\lambda_{E}\int_{0}^{\infty}L_{I_{k,e}}\exp\left(-\frac{\gamma\delta^{2}x^{\alpha_{k}}}{P_{k}\beta}\right)x\mathrm{d}x\right)$$
(29)

with

$$L_{I_{k,e}} = \exp\left(-2\pi\lambda_1 \sum_{n=1}^{S} {\binom{S}{n}} \left(\frac{\eta\gamma P_1}{SP_k}\right)^{\frac{2}{\alpha_1}} \frac{\Gamma\left(n-\frac{2}{\alpha_1}\right)\Gamma\left(-n+\frac{2}{\alpha_1}+S\right)}{\alpha_1\Gamma(S)x^{-2\alpha_k/\alpha_1}}\right).$$
(30)

**Proof.** Lemma 4 can be easily obtained following the approach in Lemma 2.

## 3.2 Secrecy spectrum efficiency

SSE, which is defined as the achievable secrecy rate over per unit of bandwidth, is proposed to measure the secrecy oriented spectrum efficiency performance in massive MIMO enabled HetNets. Based on the results derived in subsection 3.1, achievable secrecy rate is easy to be calculated and a tractable lower bound on SSE for massive MIMO enabled *K*-tier HetNets can be obtained.

**Theorem 1.** For the system model defined in Section 2, the lower bound on the SSE can be determined with

$$SSE^{L} = A_{M} \times SSE_{M}^{L} + \sum_{k=2}^{K} A_{k} \times SSE_{k} , \qquad (31)$$

where  $SSE_{M}^{L} = S \times R_{SM}^{L}$  and  $R_{SM}^{L} = \left\{R_{M}^{L} - R_{M}^{e^{*}}\right\}^{+}$  denote the lower bound on the SSE and achievable ergodic secrecy rate for macrocells, respectively. As well as  $SSE_{k} = R_{Sk}$  and  $R_{Sk} = \left\{R_{k} - R_{k}^{e^{*}}\right\}^{+}$  represent the SSE and achievable ergodic secrecy rate for the  $k^{\text{th}}$  tier small-cells, respectively.

**Proof.** The proof of Theorem 1 follows (13), (18), (21), and (28) as well as the law of total expectation.

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#### 3.3 Secrecy energy efficiency

SEE is used to evaluate the secrecy performance achieved with unit energy consumption. For sustainable operation of massive MIMO enabled HetNets, gaining theoretical understanding of SEE is as important as that of SSE. In this subsection, we present the lower bound of SEE, which is defined as the secrecy rate over the power ratio.

Theorem 2. The SEE for a massive MIMO enabled K-tier HetNet can be lower bounded as

$$\mathbf{SEE}^{L} = A_{M} \times \mathbf{SEE}_{M}^{L} + \sum_{k=2}^{K} A_{k} \times \mathbf{SEE}_{k} , \qquad (32)$$

where  $\text{SEE}_{M}^{L} = S \times R_{SM}^{L} / P_{M}^{\text{total}}$  and  $\text{SSE}_{k} = R_{Sk} / P_{k}^{\text{total}}$  denote the SEE for macrocells (lower bound) and the  $k^{\text{th}}$  tier small-cells, respectively. The each channel total power consumptions for MBS and SBS in the  $k^{\text{th}}$  tier are given by  $P_{M}^{\text{total}} = P_{1}^{0} + \frac{P_{1}}{\varepsilon_{1}} + \sum_{t=1}^{3} \left( S^{t-1} \left( \Delta_{t} + N \Lambda_{t} \right) \right)$  and

 $P_k^{\text{total}} = P_k^0 + \frac{P_k}{\varepsilon_k}$ , respectively.  $P_i^0$  and  $\varepsilon_i$  represent the static hardware power consumption

and the efficiency of the power amplifier for the *i*<sup>th</sup> tier, respectively. The parameters  $\Delta_t$  and  $\Lambda_t$  depends on the transceiver chains, precoding, coding and decoding, etc. [27-28].

**Proof.** The proof of Theorem 2 follows (13) and (19) in [7], as well as the similar lines of the proof in Theorem 1.

## 4. Numerical Results

In this section, the numerical results for the Monte Carlo simulations and the analyses are presented. The former verifies the analysis accuracy, and all the data listed in this section are averaged over 100,000 such simulations. The node locations are scattered with PPP, as described in Section 2. In this work, we consider a network with K = 2,  $\alpha_1 = 3.5$ ,  $\alpha_2 = 4$ ,  $\lambda_1 = 10^{-3}$ ,  $\lambda_2 = 2\lambda_1$ ,  $P_1 = 40P_2$ ,  $\delta^2 = 10^{-9}$ ,  $\varepsilon_1 = \varepsilon_2 = 0.38$ ,  $\Delta_1 = 4.8$ ,  $\Delta_2 = 0$ ,  $\Delta_3 = 2.08S \times 10^{-8}$ ,  $\Lambda_1 = 1$ ,  $\Lambda_2 = 9.5 \times 10^{-8}$ ,  $\Lambda_3 = 6.25 \times 10^{-8}$ ,  $P_1^0 = 4W$ ,  $P_2^0 = 13.6W$ , and  $B_2 = 1 - \eta$  [27-28]. Other simulation parameters are presented in **Table 1**. For simplicity, lower bound and Monte Carlo are denoted by (L) and (MC) in legend, respectively.

In **Fig. 1-4**, we compare the SSE for massive MIMO enabled HetNets with AN power allocation ratio, MBS's transmit antenna number, number of users served by each MBS simultaneously, and eavesdropper density, respectively. A fact clearly shown from **Fig. 1-4** is that the SSE of MBS with massive MIMO is better than that of SBS in most instance. This is because the massive MIMO can serve more users in each resource block.

**Fig. 1** shows that the SSE first improves then degrades with increasing  $\eta$ . With  $\eta$  at a value around 0.3, SEE reaches the peak. The reason is that: a) Increasing  $\eta$  increases the interference received by eavesdroppers, thus greatly degrades the receive SINR at the eavesdroppers in (9) and (11) as well as increases SSE when  $\eta$  is low. b) Increasing  $\eta$  also degrades the useful signal received by the MBS user, thus degrades the  $R_M^L$  in (13) and degrades the SSE<sub>M</sub><sup>L</sup> when  $\eta$  is high. c) The SSE for SBS, denoted by SSE<sub>2</sub>, does not substantially impact the SSE, since  $A_M$  is larger than  $A_k$  in this scenario. Moreover, SSE<sub>2</sub>

improves with increasing  $\eta$  as the useful signal received by the SBS user would not be degraded by the  $\eta$  increasing.



Fig. 1. SSE for different AN power allocation ratio





Parameters	N	S	η	$\lambda_E$
Fig.1 & 5	150	10	various	$3\lambda_1$
Fig.2 & 6	various	10	0.3	$3\lambda_1$
Fig.3 & 7	200	various	0.3	$3\lambda_1$
Fig.4 & 8	200	10	0.3	various
Fig.9	various	various	various	$3\lambda_1$

Table 1. Simulation parameters

For each curve in Fig. 2, it can be observed that both SSE and  $SSE_2$  improves with increasing MBS's transmit antenna number. Of course, increasing N can significantly improve

radio spectral efficiency for MBS. It can also decreases the average distance between the SBS user and its serving BS by reducing the  $A_k$  in (27), thus increases both the  $R_{Sk}$  and SSE<sub>2</sub> in (31).

By presenting the SSE versus the number of users served by each MBS in **Fig. 3**, we found that the SSE does not grow monotonically with *S* and reaches the peak value with *S* at a value around 20. It's partly because the SSE<sub>2</sub> degrades with increasing *S* as shown in **Fig. 3**. For large *S*, the  $R_{SM}^L$  is more influential than *S* to  $SSE_M^L = S \times R_{SM}^L$  in (31) and the  $R_{SM}^L$  degrades with increasing *S*.

**Fig. 4** plots the SSE versus the eavesdropper density. As eavesdropper density increases, the SSE accelerates to drop since both  $R_M^{e^*}$  in (18) and  $R_k^{e^*}$  in (28) increases.



Fig. 3. SSE for different number of users served by each MBS simultaneously



Fig. 4. SSE for different eavesdropper density



Fig. 6. SEE for different MBS's transmit antenna number

In **Fig. 5-8**, we compare the SEE for massive MIMO enabled HetNets with AN power allocation ratio, MBS's transmit antenna number, number of users served by each MBS simultaneously, and eavesdropper density, respectively. Most of the curves for SEE and the corresponding curves for SSE follow the same trend. Therefore, to avoid repetition, we only focus on the differences.

As shown in **Fig. 5-8**, the SEE for MBS, denoted by  $SEE_1$ , is lower than the SEE for the whole HetNet, which means the SEE of SBS is better than that of MBS. The reason is that the AN generated by MBS will deteriorate the  $SINR_k^{e^*}$  in (11) without any energy consumption on SBS.



Fig. 7. SEE for different number of users served by each MBS simultaneously



Fig. 8. SEE for different eavesdropper density

Different from SSE, as shown in **Fig. 6**, SEE does not grow monotonically with MBS's transmit antenna number and reaches the peak value at N = 200. Because increasing transmit antenna number not only improve the channel gain, but also increase the  $P_M^{\text{total}}$  in (32), which means higher energy consumption.



Fig. 9. SEE versus SSE for various values of different parameters

To balance the tradeoff between SSE and SEE, the AN power allocation ratio  $\eta$ , the number of users served by each MBS *S*, and the MBS's transmit antenna number *N* should be optimized. In order to illustrate the relation of SSE and SEE clearly, the SEE versus SSE for various values of different parameters are presented in **Fig. 9**. As can be observed from the curves in **Fig. 9**, SSE and SEE are aligned varying the value of  $\eta$  and *S*. The optimal value of  $\eta$  and *S* that maximize these two performance metrics (SSE and SEE) are 0.3 and 20, respectively. However, the optimal value of *N* that maximizes SSE is not the one that maximizes SEE. With *N* at a value around 200, the optimal SEE for the massive MIMO enabled HetNets is achieved. When *N* is more than 200, though further increase in MBS's transmit antenna number can still improve SSE, it will deteriorate SEE and greatly increase the cost and difficulty in antenna deployment. Therefore, the optimal value of *N* is 200 for the tradeoff between SSE and SEE. These optimal values can be used to guide the massive MIMO enabled HetNets design to meet the demand of security and resource-saving.

#### 5. Conclusions

In this paper, we consider a *K*-tier massive MIMO enabled HetNet, where malicious users can cooperate to eavesdrop and the MBS adopting massive MIMO can generate AN to deteriorate the channels of eavesdroppers. Firstly, the achievable ergodic rate for MBS users, SBS users, and eavesdroppers are derived, respectively. Then, a fundamental analysis framework to evaluate the SSE and SEE performance in the networks described above is proposed based on stochastic geometry. Furthermore, the impacts of AN power allocation ratio, MBS's transmit antenna number, number of users served by each MBS, and eavesdropper density on SSE as well as SEE are examined. From the numerical results of our work, the optimal SSE-SEE tradeoff is achieved at  $\eta = 0.3$ , S = 20 and N = 200, which can be adopted for advancing the design of massive MIMO enabled HetNets.

For the future work, it is important to gain a deeper understanding into SSE-SEE tradeoff in coordinated multi-point (CoMP) HetNets. Additionally, it would be interesting to investigate the SSE and SEE with wireless backhaul or imperfect channel state information in HetNets.

## **Appendix**

#### A. Proof of Lemma 1

**Proof.** Both  $f_{|X_{M,o}|}(x)$  and  $A_M$  can be obtained following the approach in [29]. The achievable ergodic rate for a randomly selected macrocell user is denoted by  $R_M$ ,

$$R_{M} = \mathbb{E}\left\{\log_{2}\left(1 + \mathrm{SINR}_{M}\right)\right\} \ge \log_{2}\left(1 + \frac{1}{\mathbb{E}\left\{\left(\mathrm{SINR}_{M}\right)^{-1}\right\}}\right) = R_{M}^{L}, \qquad (A1)$$

and

$$E\left\{\left(SINR_{M}\right)^{-1}\right\} = E\left\{\frac{S\left(I_{M}+\delta^{2}\right)}{\left(1-\eta\right)P_{1}h_{M,o}\beta\left|X_{M,o}\right|^{-\alpha_{1}}}\right\}$$

$$\stackrel{(a)}{\approx} \frac{E\left\{S\left(I_{M}+\delta^{2}\right)\left|X_{M,o}\right|^{\alpha_{1}}\right\}}{\left(1-\eta\right)P_{1}\left(N-S+1\right)\beta}$$

$$= \frac{S\int_{0}^{\infty}\left(E\left\{I_{M}\right\}+\delta^{2}\right)x^{\alpha_{1}}f_{|X_{M,o}|}(x)dx}{\left(1-\eta\right)P_{1}\left(N-S+1\right)\beta}.$$
(A2)

where step (a) follows from the law of large numbers and  $E\{I_M\} = E\{I_{M,M}\} + E\{I_{M,AN}\} + E\{I_{M,S}\}$ .  $E\{I_{M,M}\}$  can be evaluated as

$$E\{I_{M,M}\} = E\{\sum_{l \in \Phi_{1} \setminus o_{M}} \frac{(1-\eta)P_{1}}{S} h_{M,l}\beta |X_{M,l}|^{-\alpha_{1}}\}$$

$$\stackrel{(b)}{=} 2\pi\lambda_{1}E\{h_{M,l}\}\int_{x}^{\infty} \frac{(1-\eta)P_{1}}{S}\beta y^{1-\alpha_{1}}dy$$

$$= 2\pi\lambda_{1}(1-\eta)P_{1}\beta\int_{x}^{\infty} y^{1-\alpha_{1}}dy$$

$$= \frac{2\pi\lambda_{1}(1-\eta)P_{1}\beta x^{2-\alpha_{1}}}{(\alpha_{1}-2)},$$
(A3)

where step (b) follows from Campbell's theorem [30] and  $|X_{M,I}|$  is rewritten by y for ease of exposition. Similarly,  $E\{I_{M,AN}\}$  can be obtained as

$$E\{I_{M,AN}\} = E\left\{\sum_{l \in \Phi_1 \setminus \sigma_M} \frac{\eta P_1}{S} h_{AN,l} \beta \left| X_{M,l} \right|^{-\alpha_1}\right\}$$
$$= 2\pi \lambda_1 E\{h_{AN,l}\} \int_x^{\infty} \frac{\eta P_1}{S} \beta y^{1-\alpha_1} dy$$
$$= \frac{2\pi \lambda_1 \eta P_1 \beta x^{2-\alpha_1}}{(\alpha_1 - 2)},$$
(A4)

and  $E\{I_{M,S}\}$  can be derived as

$$E\left\{I_{M,S}\right\} = E\left\{\sum_{i=2}^{K} \sum_{j \in \Phi_{i}} P_{i}h_{i,j}\beta \left|X_{i,j}\right|^{-\alpha_{i}}\right\}$$

$$= \sum_{i=2}^{K} E\left\{\sum_{j \in \Phi_{i}} P_{i}h_{i,j}\beta \left|X_{i,j}\right|^{-\alpha_{i}}\right\}$$

$$= \sum_{i=2}^{K} 2\pi\lambda_{i} \int_{D_{i}^{M}(x)}^{\infty} P_{i}\beta y^{1-\alpha_{2}} dy$$

$$= \sum_{i=2}^{K} \frac{2\pi\lambda_{i} P_{i}\beta \left(D_{i}^{M}(x)\right)^{2-\alpha_{i}}}{\alpha_{i}-2},$$
(A5)

where  $D_i^M(x) = \left(\frac{P_i B_i S x^{\alpha_1}}{(N-S+1)P_1}\right)^{\frac{1}{\alpha_i}}$  is the closest distance between the interfering BS in the *i*<sup>th</sup>

tier and the selected macrocell user.

Substituting (A2), (A3), (A4) and (A5) into (A1), we can arrive at the final result.

## B. Proof of Lemma 2

**Proof.** When the typical user is connected with a MBS, the CDF of the receive SINR at the most malicious eavesdropper is derived as

$$R_{M}^{e^{*}} = E\left\{\log_{2}\left(1 + SINR_{M}^{e^{*}}\right)\right\} = \frac{1}{\ln 2} \int_{0}^{\infty} \frac{1 - F_{SINR_{M}^{e^{*}}}(\gamma)}{1 + \gamma} d\gamma, \qquad (B1)$$

where

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$$\begin{aligned} \mathbf{F}_{\mathrm{SINR}_{M}^{e^{*}}}(\gamma) &= \mathbf{P}\left(\max_{e \in \Phi_{E}} \left\{ \frac{(1-\eta)P_{1}h_{M,e}\beta \left|X_{M,e}\right|^{-\alpha_{1}}}{S\left(I_{M,e}+\delta^{2}\right)} \right\} \leq \gamma \right) \\ &= \prod_{e \in \Phi_{E}} \mathbf{P}\left(h_{M,e} \leq \frac{\gamma S\left(I_{M,e}+\delta^{2}\right) \left|X_{M,e}\right|^{\alpha_{1}}}{(1-\eta)P_{1}h_{M,e}\beta} \right) \\ &= \mathbf{E}_{\left|X_{M,e}\right|} \left\{\prod_{e \in \Phi_{E}} \mathbf{P}\left(h_{M,e} \leq \frac{\gamma S\left(I_{M,e}+\delta^{2}\right) x^{\alpha_{1}}}{(1-\eta)P_{1}h_{M,e}\beta} \right) \right\} \end{aligned} \tag{B2}$$
$$\begin{aligned} &= \exp\left(-2\pi\lambda_{E}\int_{0}^{\infty} \mathbf{E}\left\{\exp\left(-\frac{\gamma S\left(I_{M,e}+\delta^{2}\right) x^{\alpha_{1}}}{(1-\eta)P_{1}\beta}\right)\right\} x dx\right). \end{aligned}$$

where step (c) follows from the probability generating functional (PGFL) of PPP [31], and

$$\mathbf{E}\left\{\exp\left(-\frac{\gamma S\left(I_{M,e}+\delta^{2}\right)x^{\alpha_{1}}}{(1-\eta)P_{1}\beta}\right)\right\}=\mathbf{E}\left\{\exp\left(-\frac{\gamma SI_{M,e}x^{\alpha_{1}}}{(1-\eta)P_{1}\beta}\right)\right\}\exp\left(-\frac{\gamma S\delta^{2}x^{\alpha_{1}}}{(1-\eta)P_{1}\beta}\right).$$
(B3)

 $\mathbb{E}\left\{\exp\left(-\gamma SI_{M,e}x^{\alpha_{1}}/\left(\left(1-\eta\right)P_{1}\beta\right)\right)\right\} \text{ can be obtained as}$ 

$$E\left\{\exp\left(-\frac{\gamma SI_{M,e} x^{\alpha_{1}}}{(1-\eta)P_{1}\beta}\right)\right\} = E\left\{\exp\left(-\frac{\gamma \eta x^{\alpha_{1}}}{1-\eta} \sum_{l \in \Phi_{1} \setminus o_{M}} h_{e,l} \left|X_{e,l}\right|^{-\alpha_{1}}\right)\right\}$$
$$= E_{y_{e}}\left\{\prod_{l \in \Phi_{1} \setminus o_{M}} E_{h_{e,l}}\left\{\exp\left(-\frac{\gamma \eta x^{\alpha_{1}}}{1-\eta}h_{e,l} y_{e}^{-\alpha_{1}}\right)\right\}\right\}$$
$$= \exp\left(-2\pi\lambda_{1} \int_{0}^{\infty} \left(1-E_{h_{e,l}}\left\{\exp\left(-\frac{\gamma \eta x^{\alpha_{1}}}{1-\eta}h_{e,l} y_{e}^{-\alpha_{1}}\right)\right\}\right\}\right)y_{e} dy_{e}\right),$$
(B4)

where  $|X_{e,l}|$  is rewritten by  $y_e$  for ease of notation, and  $E\left\{\exp\left(-\gamma\eta x^{\alpha_1}h_{e,l}y_e^{-\alpha_1}/(1-\eta)\right)\right\}$  can be derived as

$$E_{h_{e,l}}\left\{ \exp\left(-\frac{\gamma\eta x^{\alpha_{1}}}{1-\eta}h_{e,l}y_{e}^{-\alpha_{1}}\right)\right\} = \int_{0}^{\infty} \exp\left(-\frac{\gamma\eta x^{\alpha_{1}}}{1-\eta}h_{e,l}y_{e}^{-\alpha_{1}}\right)\frac{h_{e,l}^{S-1}e^{-h_{l,e}}}{(S-1)!}dh_{l,e} \\
 = \left(1+\frac{\gamma\eta x^{\alpha_{1}}y_{e}^{-\alpha_{1}}}{1-\eta}\right)^{-S}.$$
(B5)

Furthermore,

$$\begin{split} &\int_{0}^{\infty} \left( 1 - \left( 1 + \frac{\gamma \eta x^{\alpha_{1}} y_{e}^{-\alpha_{1}}}{1 - \eta} \right)^{-S} \right) y_{e} dy_{e} \\ &= \int_{0}^{\infty} \sum_{n=1}^{S} \binom{S}{n} \left( \frac{\gamma \eta x^{\alpha_{1}} y_{e}^{-\alpha_{1}}}{1 - \eta} \right)^{n} \left( 1 + \frac{\gamma \eta x^{\alpha_{1}} y_{e}^{-\alpha_{1}}}{1 - \eta} \right)^{-S} y_{e} dy_{e} \\ &= \frac{1}{\alpha_{1}} \int_{0}^{\infty} \sum_{n=1}^{S} \binom{S}{n} \left( \frac{\gamma \eta x^{\alpha_{1}} t}{1 - \eta} \right)^{n} \left( 1 + \frac{\gamma \eta x^{\alpha_{1}} t}{1 - \eta} \right)^{-S} t^{-(\alpha_{1} + 2)/\alpha_{1}} dt \\ &= \sum_{n=1}^{S} \binom{S}{n} \frac{\left( \frac{\gamma \eta x^{\alpha_{1}}}{1 - \eta} \right)^{2/\alpha_{1}}}{\alpha_{1} \Gamma(S)} \Gamma \left( -n + \frac{2}{\alpha_{1}} + S \right)}, \end{split}$$
(B6)

where  $t = y_e^{-\alpha_1}$ .

With the above results, we can arrive at the final result.

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